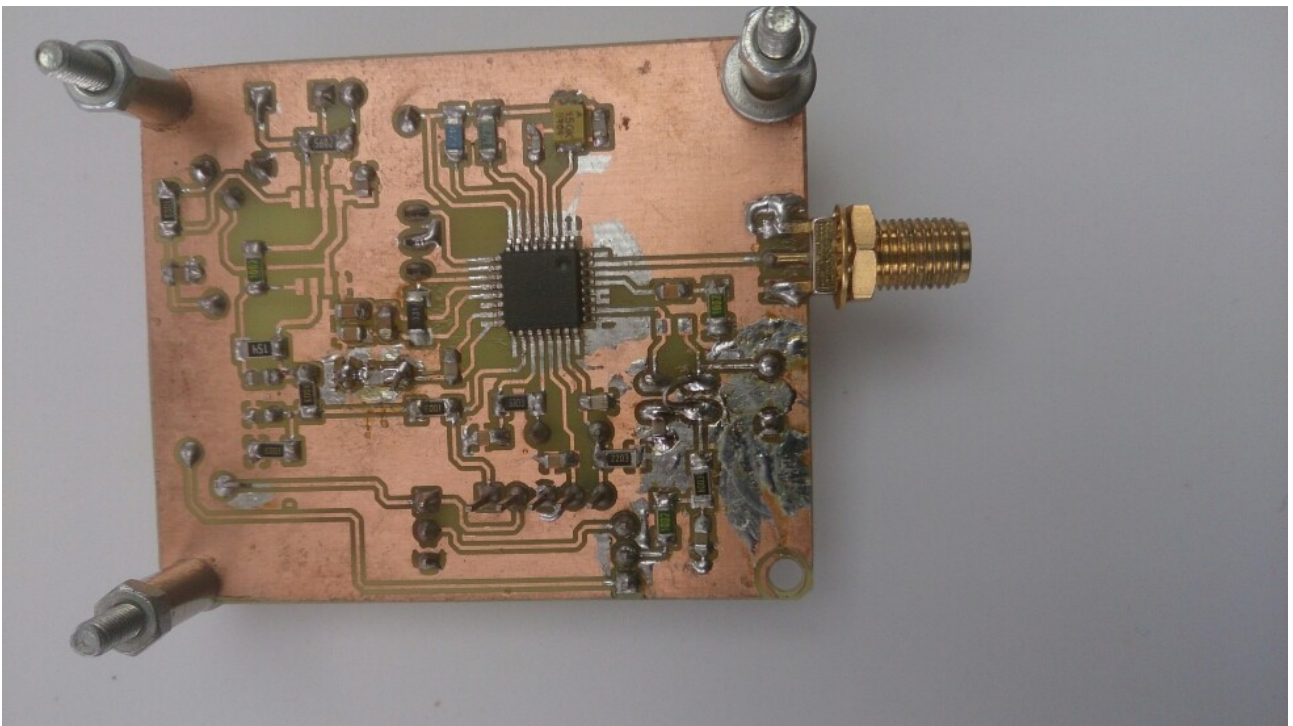


Récepteur 406Mhz FSK et CAF de F1LVT TH71101 de chez MELEXIS



F 5 L E B – Denis
f5leb@club-internet.fr

Modules de réception UHF des balises de détresse à l'attention des ADRASEC

L'approvisionnement des platines Wawix est arrivé à son terme, il convient donc de chercher un autre type de module de réception 406Mhz. Mon choix s'est porté sur une puce de chez Melexis le TH71101 qui est en tout point identique au composant de chez Microchip rfRxd0420 qui, malheureusement, n'est plus fabriqué, ce composant permet la réception en mode FSK.

Pour dessiner le circuit j'utilise KICAD, mais l'ancienne version ([2013.07.07-BZR4022 win full version.exe](#)), cette version se trouve sur le site de KICAD, download, windows, lien en bas de la page.

Le schéma ne pose guère de problème, les composants sont des cms, on peut trouver la puce sur internet « Melexis TH71101 » ou chez « Mouser » ou chez « DIGIKEY » ou ailleurs, chacun fera son choix, je préconise de se grouper afin d'avoir un prix d'achat pas trop élevé, pour ma part, je l'ai acheté chez « DIGIKEY » mais les frais de douane sont élevés.

Je peux vous faire parvenir le dossier zippé pour Kicad,

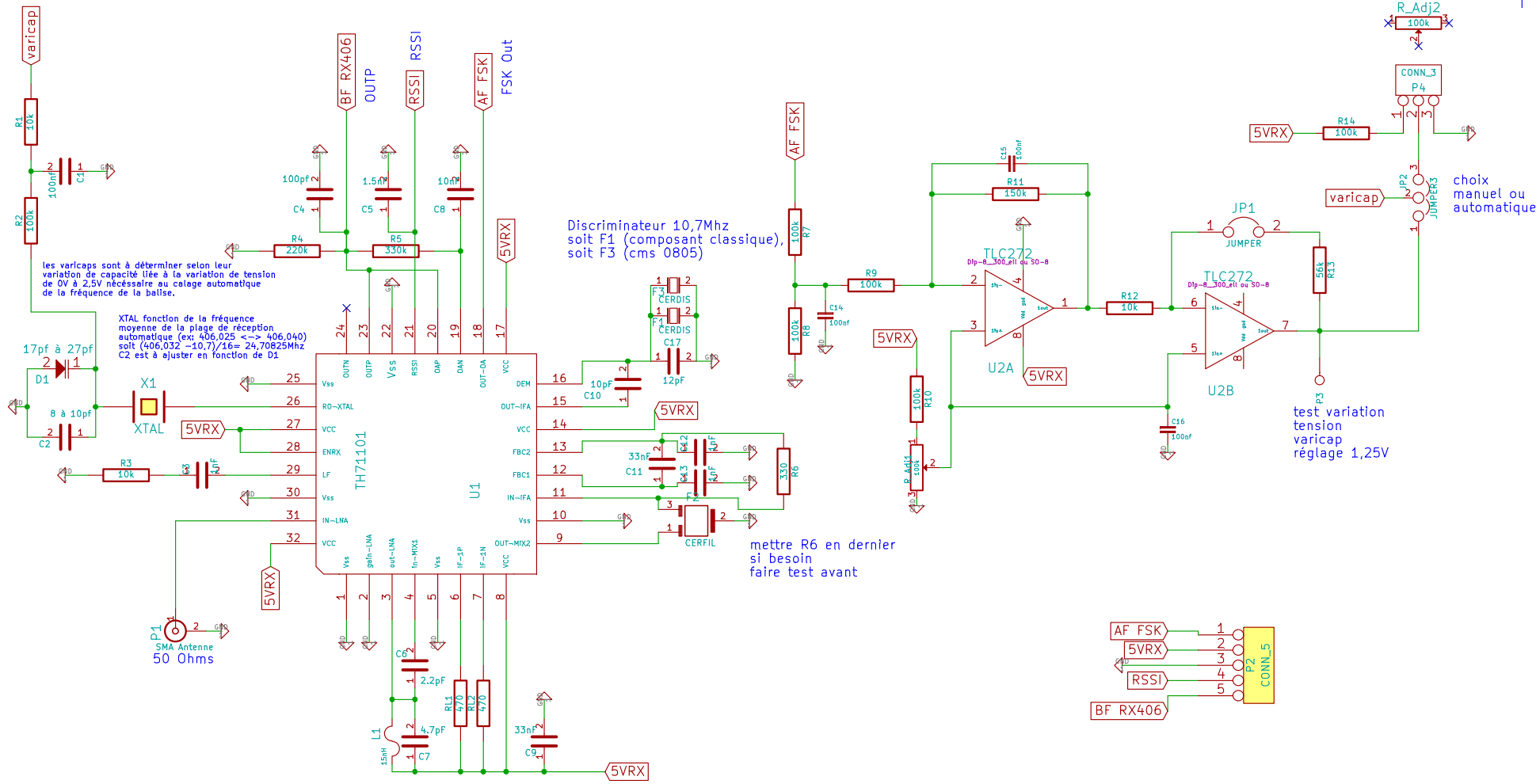
Quelques infos sur cette réalisation, j'ai rajouté au schéma d'origine le montage de Jean-Paul F1LVT sur la recherche automatique de l'émission de la balise, ainsi que la possibilité de tester la variation de fréquence manuellement. La diode varicap est de récupération, la plage de fonctionnement est de 17pF à 27pF, à laquelle on rajoute une capa de 8 à 10pF afin d'obtenir une variation de 35pF à 25pF, le quartz est taillé à 24,70825Mhz, soit une fréquence centrale de 406,032, ce qui doit permettre de balayer une bande de 406,025Mhz à 406,040Mhz. Pour tester la valeur de la varicap, je joins le schéma du montage test.

En ce qui concerne le discriminateur 10,7Mhz, j'ai prévu deux type d'emplacement, soit un CMS (empreinte = SM2010), soit un composant ordinaire, auquel il faudra rajouter un condensateur de 10 à 12pF en // sur le discri comme prévu sur le plan.

Cordiales 73

F5LEB, Denis

f5leb@club-internet.fr



les varicaps sont à déterminer selon leur variation de capacité liée à la variation de tension de 0V à 2,5V nécessaire au calage automatique de la fréquence de la balise.

XTAL fonction de la fréquence moyenne de la plage de réception automatique (ex: 406,025 <-> 406,040) soit $(406,032 - 10,7)/16 = 24,70825\text{Mhz}$ C2 est à ajuster en fonction de D1

Discriminateur 10,7Mhz soit F1 (composant classique), soit F3 (cms 0805)

mettre R6 en dernier si besoin faire test avant

test variation tension varicap réglage 1,25V

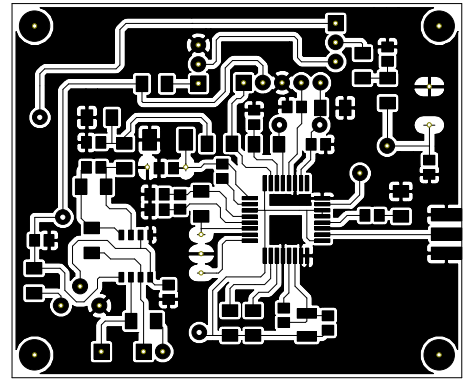
choix manuel ou automatique

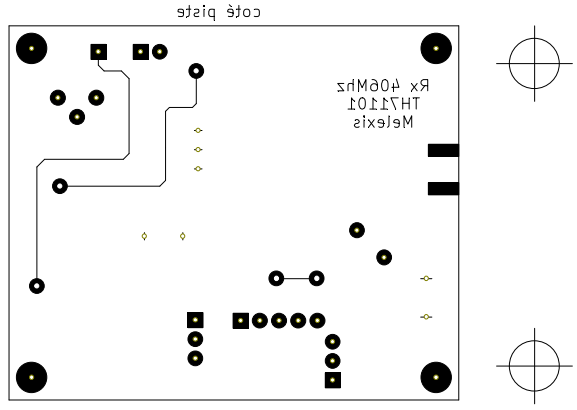
CERFIL = SFECF10M7HA00 B3db = 180Hz
 CERDIS = CDSCB10M7GA135 discriminateur
 SAWFIL = (433,92 Mhz) SAFCC433MBLOX00 B3db = 840khz
 XTAL = $(R_x - 10,7)/16$
 Vcc = 5v

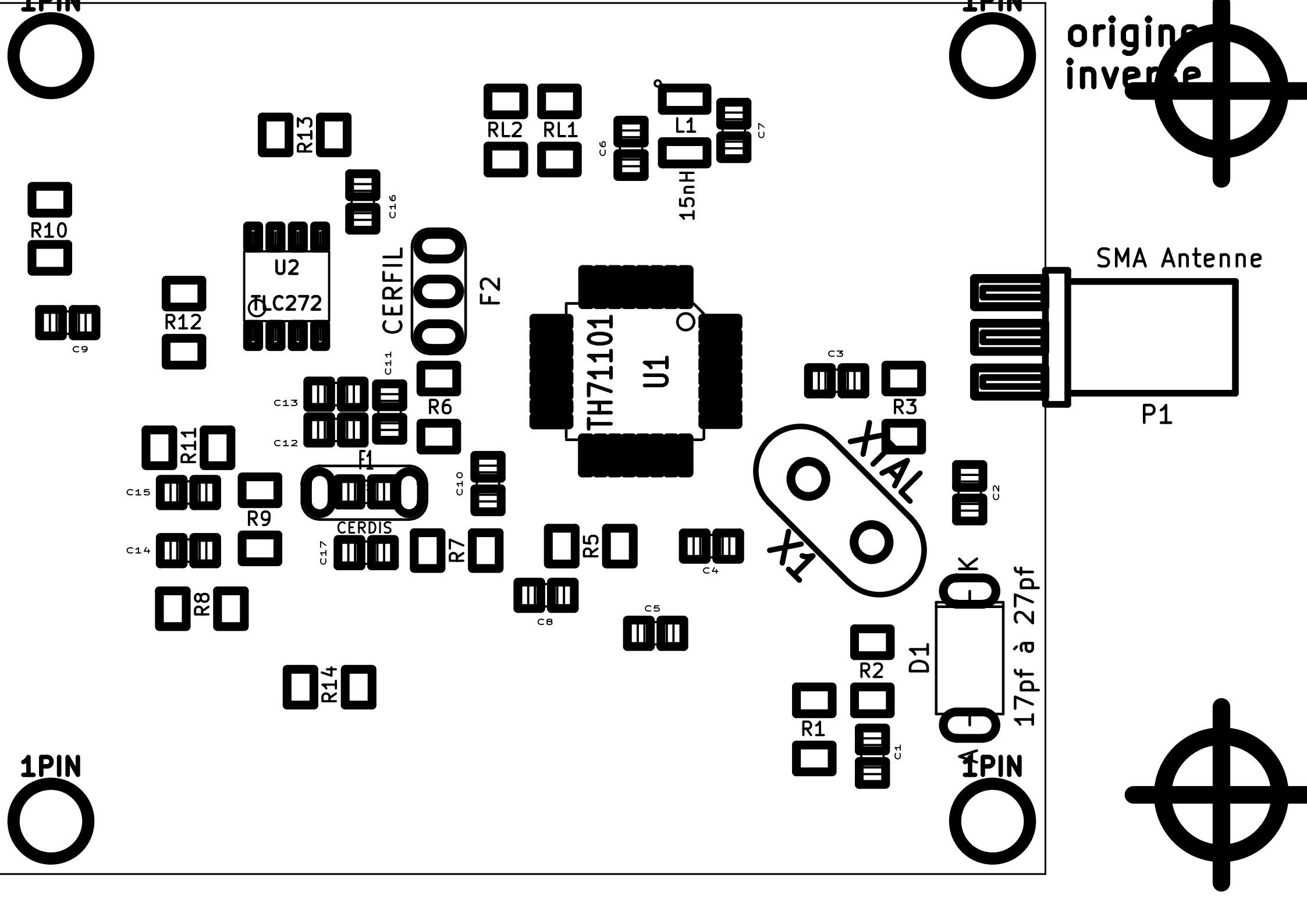
Récepteur FSK standard pour 406 Mhz
 selon schéma de chez MELEXIS
 Réception balises de détresse

ADRASEC		
File: Rx_TH71101_caf_F1LVT.sch		
Sheet: /		
Title: Rx 406Mhz FSK avec CAF de F1LVT		
Size: A4	Date: 30 jul 2018	Rev: F5LEB
KiCad E.D.A. eschema (2013-07-07 BZR 4022)-stable		Id: 1/1

cofe components







eeschema (2013-07-07 BZR 4022)-stable >> Creation date: 30/07/2018 20:44:57

#Cmp (ordre = Référence)

C1	100nf
C2	8 à 10pf
C3	1nF
C4	100pf
C5	1.5nF
C6	2.2pF
C7	4.7pF
C8	10nF
C9	33nF
C10	10pF
C11	33nF
C12	1nF
C13	1nF
C14	100nf
C15	100nf
C16	100nf
C17	12pF
D1	17pf à 27pf
F1	CERDIS
F2	CERFIL
F3	CERDIS
JP1	JUMPER
JP2	JUMPER3
L1	15nH
P1	SMA Antenne
P2	CONN_5
P3	CONN_1
P4	CONN_3
R1	10k
R2	100k
R3	10k
R4	220k
R5	330k
R6	330
R7	100k
R8	100k
R9	100k
R10	100k
R11	150k
R12	10k
R13	56k
R14	100k
R_Adj1	100k
R_Adj2	100k
RL1	470
RL2	470
U1	TH71101
U2	TLC272
X1	XTAL

#End Cmp

Rx_TH71101_caf_F1LVT.lst

```

#Cmp ( ordre = Valeur )
| 2.2pF      C6
| 100pf      C4
| 1nF        C3
| 100nf      C14
| 100k       R_Adj1
| 4.7pF      C7
| 10pF       C10
| 1nF        C12
| 1nF        C13
| 1.5nF      C5
| 10nF       C8
| 33nF       C9
| 33nF       C11
| 100k       R_Adj2
| 12pF       C17
| 100nf      C1
| 100nf      C15
| 100nf      C16
| 8 à 10pf   C2
| 17pf à 27pf D1
| CERDIS     F1
| CERFIL     F2
| CERDIS     F3
| JUMPER     JP1
| JUMPER3    JP2
| 15nH       L1
| SMA Antenne P1
| CONN_5     P2
| CONN_1     P3
| CONN_3     P4
| 330        R6
| 10k        R1
| 10k        R3
| 10k        R12
| 56k        R13
| 100k       R2
| 100k       R7
| 100k       R8
| 100k       R9
| 100k       R10
| 100k       R14
| 150k       R11
| 220k       R4
| 330k       R5
| 470        RL1
| 470        RL2
| TLC272     U2
| TH71101    U1
| XTAL       X1
#End Cmp

```


Rx_TH71101_caf_F1LVT.lst

#End List

Features

- Single-conversion superhet architecture for low external component count
- FSK demodulation with phase-coincidence demodulator
- Low current consumption in active mode and very low standby current
- Switchable LNA gain for improved dynamic range
- RSSI allows signal strength indication and ASK detection
- 32-pin Low profile Quad Flat Package (LQFP)

Ordering Code

Product Code	Temperature Code	Package Code	Option Code	Packing Form Code
TH71101	E	NE	CAA-000	RE
TH71101	E	NE	CAA-000	TR

Legend:

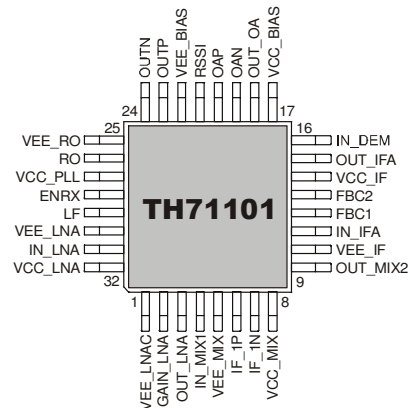
Temperature Code: E for Temperature Range -40°C to 85°C
 Package Code: NE for LQFP
 Packing Form: RE for Reel, TR for Tray

Ordering example: TH71101ENE-CAA-000-RE

Application Examples

- General digital data transmission
- Tire Pressure Monitoring Systems (TPMS)
- Remote Keyless Entry (RKE)
- Wireless access control
- Alarm and security systems
- Garage door openers
- Remote Controls
- Home and building automation
- Low-power telemetry systems

Pin Description



General Description

The TH71101 FSK/ASK single-conversion superheterodyne receiver IC is designed for applications in the European 433MHz industrial-scientific-medical (ISM) band, according to the EN 300 220 telecommunications standard. It can also be used for any other system with carrier frequencies ranging from 300MHz to 450MHz (e.g. for applications according to FCC part 15 and ARIB STD-T67).

Document Content

1	Theory of Operation	3
1.1	General	3
1.2	Technical Data Overview	3
1.3	Block Diagram.....	4
1.4	Mode Configurations	4
1.5	LNA GAIN Control.....	4
1.6	Frequency Planning	4
1.6.1	Selected Frequency Plans	5
1.6.2	Maximum Frequency Coverage.....	5
2	Pin Definitions and Descriptions	6
3	Technical Data	9
3.1	Absolute Maximum Ratings.....	9
3.2	Normal Operating Conditions	9
3.3	Crystal Parameters	9
3.4	DC Characteristics	10
3.5	AC System Characteristics.....	11
4	Test Circuits	12
4.1	Standard FSK Reception.....	12
4.1.1	Standard FSK Application Circuit.....	12
4.1.2	Standard FSK Component List	13
4.2	Narrow Band FSK Reception	14
4.2.1	Narrow Band FSK Application Circuit	14
4.2.2	Narrow Band FSK Component List.....	15
4.3	ASK Reception.....	16
4.3.1	ASK Application Circuit	16
4.3.2	ASK Component List.....	17
5	Package Description	18
5.1	Soldering Information.....	18
6	Standard information regarding manufacturability of Melexis products with different soldering processes	19
7	ESD Precautions	19
8	Disclaimer	20

1 Theory of Operation

1.1 General

With the TH71101 receiver chip, various circuit configurations can be arranged in order to meet a number of different customer requirements. For FSK reception the IF tank used in the phase coincidence demodulator can be constituted by an external ceramic discriminator. In ASK configuration, the RSSI signal is fed to an ASK detector, which is constituted by the operational amplifier.

A double-conversion variant, called TH71102, is also available. This receiver IC allows a higher degree of image rejection, achieved in conjunction with an RF front-end filter. Both RXICs have the same die. At the TH71102, the second mixer (MIX2) is used to down-convert the first IF (IF1) to the second IF (IF2). At the TH71101, MIX2 operates as an amplifier.

Efficient RF front-end filtering is realized by using a SAW, ceramic or helix filter in front of the LNA and by adding an LC filter at the LNA output.

The TH71101 receiver IC consists of the following building blocks:

- PLL synthesizer (PLL SYNTH) for generation of the local oscillator signal LO, parts of the PLL SYNTH are: the high-frequency VCO1, the feedback divider DIV_16, a phase-frequency detector (PFD) with charge pump (CP) and a crystal-based reference oscillator (RO)
- Low-noise amplifier (LNA) for high-sensitivity RF signal reception
- First mixer (MIX1) for down-conversion of the RF signal to the IF
- IF pre amplifier which is a mixer cell (MIX2) that operates as an amplifier
- IF amplifier (IFA) to amplify and limit the IF signal and for RSSI generation
- Phase coincidence demodulator (DEMOD) with third mixer (MIX3) to demodulate the IF signal
- Operational amplifier (OA) for data slicing, filtering and ASK detection
- Bias circuitry for bandgap biasing and circuit shutdown

1.2 Technical Data Overview

- | | |
|---|--|
| <input type="checkbox"/> Input frequency range: 300 to 450 MHz | <input type="checkbox"/> Range of IF: 400 kHz to 22 MHz |
| <input type="checkbox"/> Power supply range: 2.3 to 5.5 V @ ASK | <input type="checkbox"/> Maximum input level: -10 dBm @ ASK |
| <input type="checkbox"/> Temperature range: -40 to +85 °C | <input type="checkbox"/> 0 dBm @ FSK |
| <input type="checkbox"/> Standby current: 50 nA | <input type="checkbox"/> Image rejection: > 45 dB (e.g. with 433.92 MHz SAW front-end filter and at 10.7 MHz IF) |
| <input type="checkbox"/> Operating current: 6.5 mA @ low gain | <input type="checkbox"/> Spurious emission: < -70 dBm |
| <input type="checkbox"/> 8.2 mA @ high gain | <input type="checkbox"/> Input frequency acceptance range: up to ±100 kHz |
| <input type="checkbox"/> Sensitivity: -113 dBm @ ASK 1) | <input type="checkbox"/> RSSI range: 70 dB |
| <input type="checkbox"/> -107 dBm @ FSK 2) | <input type="checkbox"/> FSK deviation range: ±2.5 kHz to ±80 kHz |
| <input type="checkbox"/> Maximum data rate: 260 kbps NRZ @ ASK | <input type="checkbox"/> |
| <input type="checkbox"/> 180 kbps NRZ @ FSK | <input type="checkbox"/> |

- 1) at 4 kbps NRZ, BER = $3 \cdot 10^{-3}$, 180 kHz IF filter BW, without SAW front-end-filter loss
- 2) at 4 kbps NRZ, BER = $3 \cdot 10^{-3}$, ± 20 kHz FSK deviation, 180 kHz IF filter BW, without SAW front-end-filter loss

1.3 Block Diagram

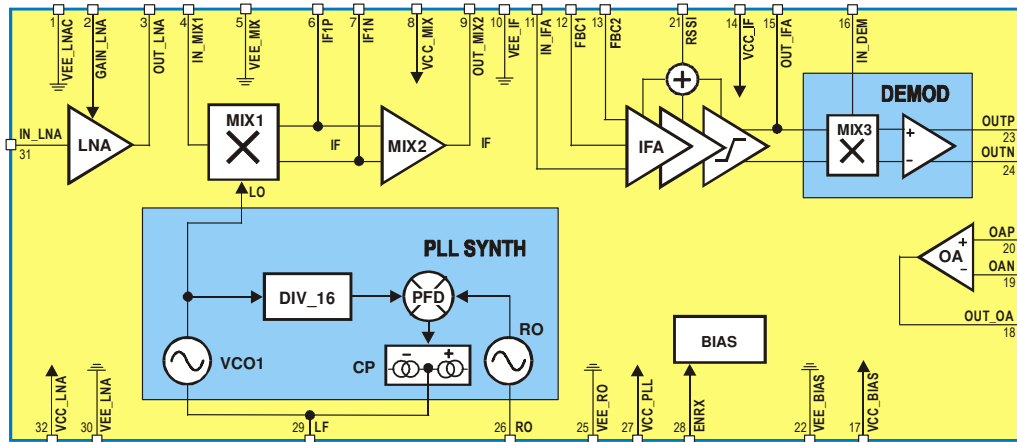


Fig. 1: TH71101 block diagram

1.4 Mode Configurations

ENRX	Mode	Description
0	RX standby	RX disabled
1	RX active	RX enable

Note: ENRX are pulled down internally

1.5 LNA GAIN Control

V_{GAIN_LNA}	Mode	Description
< 0.8 V	HIGH GAIN	LNA set to high gain
> 1.4 V	LOW GAIN	LNA set to low gain

Note: hysteresis between gain modes to ensure stability

1.6 Frequency Planning

Frequency planning is straightforward for single-conversion applications because there is only one IF that can be chosen, and then the only possible choice is low-side or high-side injection of the LO signal (which is now the one and only LO signal in the receiver).

The receiver's single-conversion architecture requires careful frequency planning. Besides the desired RF input signal, there are a number of spurious signals that may cause an undesired response at the output. Among them is the image of the RF signal that must be suppressed by the RF front-end filter.

By using the internal PLL synthesizer of the TH71101 with the fixed feedback divider ratio of $N = 16$ (DIV_16), two types of down-conversion are possible: low-side injection of LO and high-side injection of LO. The following table summarizes some equations that are useful to calculate the crystal reference frequency (REF) and the LO frequency, for a given RF and IF.

Injection type	low	high
REF	$(RF - IF)/16$	$(RF + IF)/16$
LO	$16 \cdot REF$	$16 \cdot REF$
IF	$RF - LO$	$LO - RF$
RF image	$RF - 2IF$	$RF + 2IF$

1.6.1 Selected Frequency Plans

The following table depicts crystal, LO and image signals considering the examples of 315 MHz and 433.92 MHz RF reception at IF = 10.7 MHz.

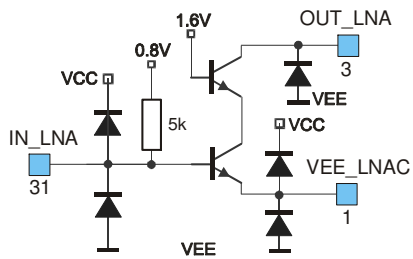
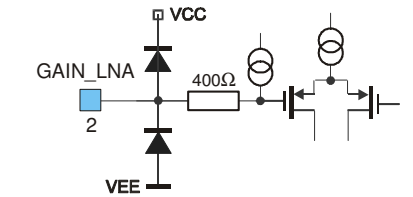
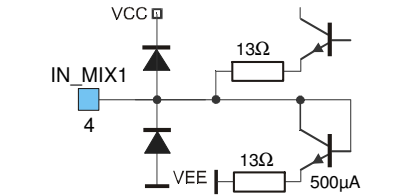
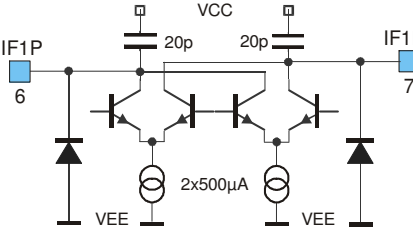
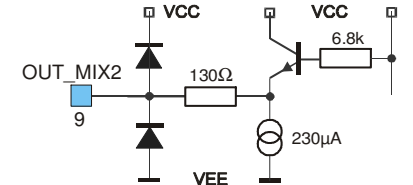
Signal type	RF = 315 MHz	RF = 315 MHz	RF = 433.92 MHz	RF = 433.92 MHz
Injection type	low	high	low	high
REF / MHz	19.01875	20.35625	26.45125	27.78875
LO / MHz	304.3	325.7	423.22	444.62
RF image / MHz	293.6	336.4	412.52	455.32

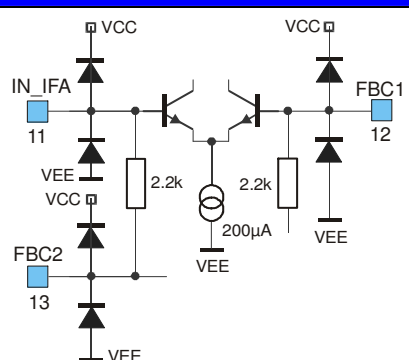
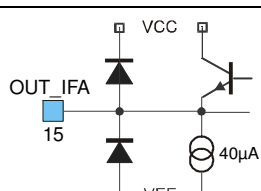
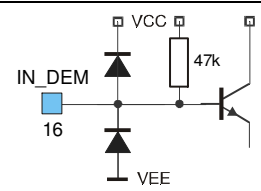
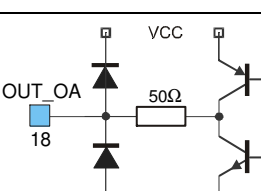
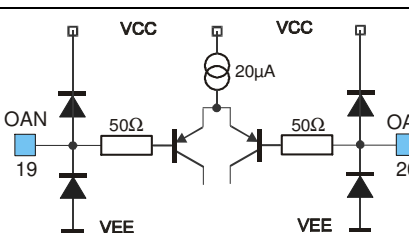
The selection of the reference crystal frequency is based on some assumptions. As for example: the image frequency should not be in a radio band where strong interfering signals might occur (because they could represent parasitic receiving signals), the LO signal should be in the range of 300 MHz to 450 MHz (because this is the optimum frequency range of the VCO1). Furthermore the IF should be as high as possible to achieve highest RF image rejection. The columns in bold depict the selected frequency plans to receive at 315 MHz and 433.92 MHz, respectively.

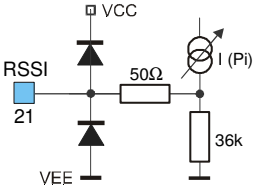
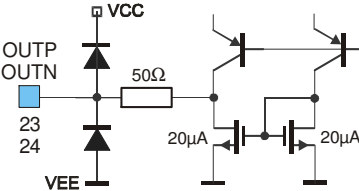
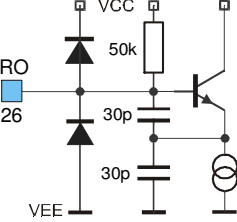
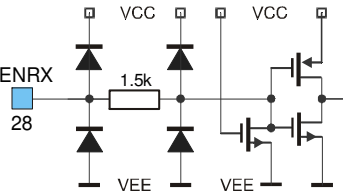
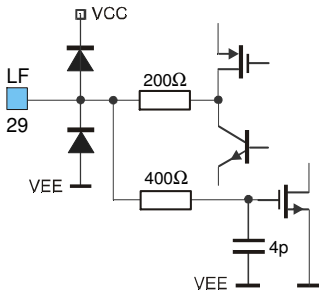
1.6.2 Maximum Frequency Coverage

Parameter	f_{min}	f_{max}
Injection type	high	low
RF / MHz	289.3	460.7
REF / MHz	18.75	28.125
LO / MHz	300	450
IF / MHz	10.7	10.7

2 Pin Definitions and Descriptions

Pin No.	Name	I/O Type	Functional Schematic	Description
3	OUT_LNA	analog output		LNA open-collector output, to be connected to external LC tank that resonates at RF
31	IN_LNA	analog input		LNA input, approx. 26Ω single-ended
1	VEE_LNAC	ground		ground of LNA core (cas-code)
2	GAIN_LNA	analog input		LNA gain control (input with hysteresis) RX standby: no pull-up RX active: pull-up
4	IN_MIX1	analog input		MIX1 input, approx. 33Ω single-ended
5	VEE_MIX	ground		ground of MIX1 and MIX2
6	IF1P	analog I/O		open-collector output, to be connected to external LC tank that resonates at first IF
7	IF1N	analog I/O		open-collector output, to be connected to external LC tank that resonates at first IF
8	VCC_MIX	supply		positive supply of MIX1 and MIX2
9	OUT_MIX2	analog output		MIX2 output, approx. 330Ω output impedance
10	VEE_IF	ground		ground of IFA and DEMOD

Pin No.	Name	I/O Type	Functional Schematic	Description
11	IN_IFA	analog input		IFA input, approx. 2.2kΩ input impedance
12	FBC1	analog I/O		to be connected to external IFA feedback capacitor
13	FBC2	analog I/O		to be connected to external IFA feedback capacitor
14	VCC_IF	supply		positive supply of IFA and DEMOD
15	OUT_IFA	analog I/O		IFA output and MIX3 input (of DEMOD)
16	IN_DEM	analog input		DEMOD input, to MIX3 core
17	VCC_BIAS	supply		positive supply of general bias system and OA
18	OUT_OA	analog output		OA output, 40uA current drive capability
19	OAN	analog input		negative OA input
20	OAP	analog input		positive OA input

Pin No.	Name	I/O Type	Functional Schematic	Description
21	RSSI	analog output		RSSI output, for RSSI and ASK detection, approx. 36kΩ output impedance
22	VEE_BIAS	ground		ground of general bias system and OA
23	OUTP	analog output		FSK positive output, output impedance of 100kΩ to 300kΩ
24	OUTN	analog output		FSK negative output, output impedance of 100kΩ to 300kΩ
25	VEE_RO	ground		ground of DIV, PFD, RO and charge pump
26	RO	analog input		RO input, Colpitts type oscillator with internal feedback capacitors
27	VCC_PLL	supply		positive supply of DIV, PFD, RO and charge pump
28	ENRX	digital input		mode control input, CMOS-compatible with internal pull-down circuit
29	LF	analog I/O		charge pump output and VCO1 control input
30	VEE_LNA	ground		ground of LNA biasing
32	VCC_LNA	supply		positive supply of LNA biasing

3 Technical Data

3.1 Absolute Maximum Ratings

Parameter	Symbol	Condition / Note	Min	Max	Unit
Supply voltage	V_{CC}		0	7.0	V
Input voltage	V_{IN}		- 0.3	$V_{CC}+0.3$	V
Input RF level	P_{IRF}	@ LNA input		10	dBm
Storage temperature	T_{STG}		-40	+125	°C
Junction temperature	T_J			+150	°C
Thermal Resistance	R_{thJA}			60	K/W
Power dissipation	P_{diss}			0.1	W
Electrostatic discharge	V_{ESD1}	human body model, 3)	-1.0	+1.0	kV
	V_{ESD2}	human body model, 4)	-0.75	+0.75	

3) all pins except OUT_LNA, IF1P and IF1N

4) pin OUT_LNA, IF1P and IF1N

3.2 Normal Operating Conditions

Parameter	Symbol	Condition	Min	Max	Unit
Supply voltage	$V_{CC, FSK}$	0 °C to 85 °C	2.5	5.5	V
		-20 °C to 85 °C	2.6	5.5	
		-40 °C to 85 °C	2.7	5.5	
	$V_{CC, ASK}$	-40 °C to 85 °C	2.3	5.5	
Operating temperature	T_A		-40	+85	°C
Input low voltage (CMOS)	V_{IL}	ENRX pin		$0.3 \cdot V_{CC}$	V
Input high voltage (CMOS)	V_{IH}	ENRX pin	$0.7 \cdot V_{CC}$		V
Input frequency range	f_i		289.3	460.7	MHz
IF range	f_{IF}		0.4	22	MHz
XOSC frequency	f_{ref}	set by the crystal	18.75	28.125	MHz
VCO frequency	f_{LO}	$f_{LO} = 16 \cdot f_{ref}$	300	450	MHz
Frequency deviation	Δf		± 2.5	± 80	kHz
FSK data rate	R_{FSK}	NRZ, C15 = NIP, 5)		180	kbps
ASK data rate	R_{ASK}	NRZ, C16 = NIP, 5)		260	kbps

5) $B_{IF} = 400$ kHz, $P_{IN} = -90$ dBm

3.3 Crystal Parameters

Parameter	Symbol	Condition	Min	Max	Unit
Crystal frequency	f_0	fundamental mode, AT	18.75	28.125	MHz
Load capacitance	C_L		10	15	pF
Static capacitance	C_0			7	pF
Series resistance	R_1			50	Ω

3.4 DC Characteristics

all parameters under normal operating conditions, unless otherwise stated;
typical values at $T_A = 23\text{ }^\circ\text{C}$ and $V_{CC} = 3\text{ V}$

Parameter	Symbol	Condition	Min	Typ	Max	Unit
Operating Currents						
Standby current	I_{SBY}	ENRX=0		50	100	nA
Supply current at low gain	$I_{CC, low}$	ENRX=1 GAIN_LNA=1	4.0	6.5	12.0	mA
Supply current at high gain	$I_{CC, high}$	ENRX=1 GAIN_LNA=0	4.5	8.2	14.0	mA
Digital Pin Characteristics						
Input low voltage CMOS	V_{IL}	ENRX pin	-0.3		$0.3 \cdot V_{CC}$	V
Input high voltage CMOS	V_{IH}	ENRX pin	$0.7 \cdot V_{CC}$		$V_{CC} + 0.3$	V
Pull down current ENRX pin	I_{PDEN}	ENRX=1	0.1	2	10	μA
Low level input current ENRX pin	I_{INLEN}	ENRX=0			0.05	μA
Analog Pin Characteristics						
High level input current GAIN_LNA pin	I_{INHAIN}	GAIN_LNA=1			0.05	μA
Pull up current GAIN_LNA pin active	$I_{PUGAINa}$	GAIN_LNA=0 ENRX=1	0.08	0.15	0.3	μA
Pull up current GAIN_LNA pin standby	$I_{PUGAINs}$	GAIN_LNA=0 ENRX=0			0.05	μA
High gain input voltage	V_{IHGAIN}	ENRX=1			0.7	V
Low gain input voltage	V_{ILGAIN}	ENRX=1	1.5			V
Opamp Characteristics						
Opamp input offset voltage	V_{offs}		-35		35	mV
Opamp input offset current	I_{offs}	$I_{OAP} - I_{OAN}$	-50		50	nA
Opamp input bias current	I_{bias}	$0.5 \cdot (I_{OAP} + I_{OAN})$	-150		150	nA
RSSI Characteristics						
RSSI voltage at low input level	$V_{RSSI, low}$	$P_1 = -65\text{ dBm}$, GAIN_LNA=1	0.5	1.0	1.5	V
RSSI voltage at high input level	$V_{RSSI, high}$	$P_1 = -35\text{ dBm}$, GAIN_LNA=1	1.2	1.9	2.5	V

3.5 AC System Characteristics

all parameters under normal operating conditions, unless otherwise stated;
 typical values at $T_A = 23\text{ }^\circ\text{C}$ and $V_{CC} = 3\text{ V}$,
 RF at 433.92 MHz; SAW frond-end filter loss and IF at 10.7 MHz;
 all parameters based on test circuits as shown in Fig. 2, Fig.3 and Fig. 5

Parameter	Symbol	Condition	Min	Typ	Max	Unit
Receive Characteristics						
Input sensitivity – FSK (standard)	$P_{\min, ST}$	$B_{IF} = 180\text{kHz}$, $\Delta f = \pm 20\text{kHz}$, 4kbps NRZ, $BER \leq 3 \cdot 10^{-3}$, 6)		-104		dBm
Input sensitivity – FSK (narrow band)	$P_{\min, NB}$	$B_{IF} = 30\text{kHz}$, $\Delta f = \pm 5\text{kHz}$, 4kbps NRZ, $BER \leq 3 \cdot 10^{-3}$, 6)		-108		dBm
Input sensitivity – ASK	$P_{\min, ASK}$	$B_{IF} = 180\text{kHz}$, 4kbps NRZ, $BER \leq 3 \cdot 10^{-3}$, 6)		-110		dBm
Maximum input signal – FSK	$P_{\max, FSK}$	$BER \leq 3 \cdot 10^{-3}$ GAIN_LNA = 1		0		dBm
Maximum input signal – ASK	$P_{\max, ASK}$	$BER \leq 3 \cdot 10^{-3}$ GAIN_LNA = 1		-10		dBm
Spurious emission	P_{spur}				-70	dBm
Image rejection	ΔP_{imag}			45		dB
Start-up Parameters						
Crystal start-up time	T_{XTL}	ENRX from 0 to 1			0.9	ms
Receiver start-up time	T_{RX}	ENRX from 0 to 1, depends on data slicer time constant, valid data at output			T_{XTL} + $R4 \cdot C17$	
PLL Parameters						
VCO gain	K_{VCO}			250		MHz/V
Charge pump current	I_{CP}			60		μA

6) incl. 3 dB loss of front-end SAW filter

4 Test Circuits

4.1 Standard FSK Reception

4.1.1 Standard FSK Application Circuit

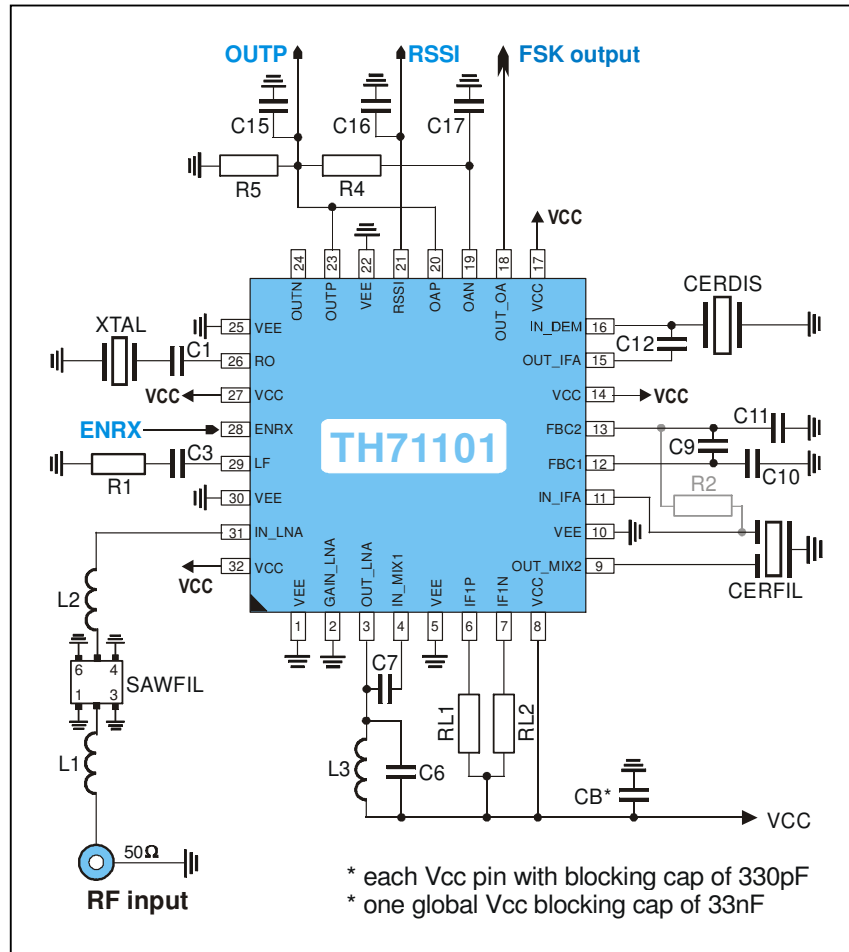


Fig. 2: Test circuit for FSK reception

Circuit Features

- Tolerates input frequency variations
- Well-suited for NRZ, Manchester and similar codes

4.1.2 Standard FSK Component List

Part	Size	Value @ 433.92 MHz	Tolerance	Description
C1	0805	27 pF	±5%	crystal series capacitor
C3	0603	1 nF	±10%	loop filter capacitor
C6	0603	4.7 pF	±5%	LNA output tank capacitor
C7	0603	2.2 pF	±5%	MIX1 input matching capacitor
C9	0603	33 nF	±10%	IFA feedback capacitor
C10	0603	1 nF	±10%	IFA feedback capacitor
C11	0603	1 nF	±10%	IFA feedback capacitor
C12	0805	10 pF	±5%	DEMODO phase-shift capacitor
C15	0805	100 pF	±5%	demodulator output low-pass capacitor, this value for data rates < 20 kbps NRZ
C16	0805	1.5 nF	±10%	RSSI output low-pass capacitor
C17	0805	10 nF	±10%	data slicer capacitor, this value for data rates > 0.8 kbps NRZ
R1	0603	10 kΩ	±5%	loop filter resistor
R2	0603	330 Ω	±5%	optional CERFIL output matching resistor
R4	0805	330 kΩ	±5%	data slicer resistor
R5	0805	220 kΩ	±5%	loading resistor
RL1	0805	470 Ω	±5%	MIX1 bias resistor
RL2	0805	470 Ω	±5%	MIX1 bias resistor
L1	0603	68 nH	±5%	SAW filter matching inductor from Würth-Elektronik (WE-KI series), or equivalent part
L2	0603	82 nH	±5%	
L3	0603	15 nH	±5%	LNA output tank inductor from Würth-Elektronik (WE-KI series), or equivalent part
XTAL	SMD 6x3.5	26.45125 MHz @ RF = 433.92 MHz	±25ppm cal. ±30ppm temp.	fundamental-mode crystal from Telcona/Horizon or equivalent part
SAWFIL	SMD 3x3	SAFCC433MBL0X00 (f ₀ = 433.92 MHz)	B _{3dB} = 840 kHz	low-loss SAW filter from Murata, or equivalent part
CERFIL	SMD 3.45x3.1	SFECF10M7HA00	B _{3dB} = 180 kHz	ceramic filter from Murata, or equivalent part
CERDIS	SMD 4.5x2	CDSCB10M7GA135		ceramic discriminator from Murata, or equivalent part

- For component values for other frequencies, please refer to the EVB descriptions

4.2 Narrow Band FSK Reception

4.2.1 Narrow Band FSK Application Circuit

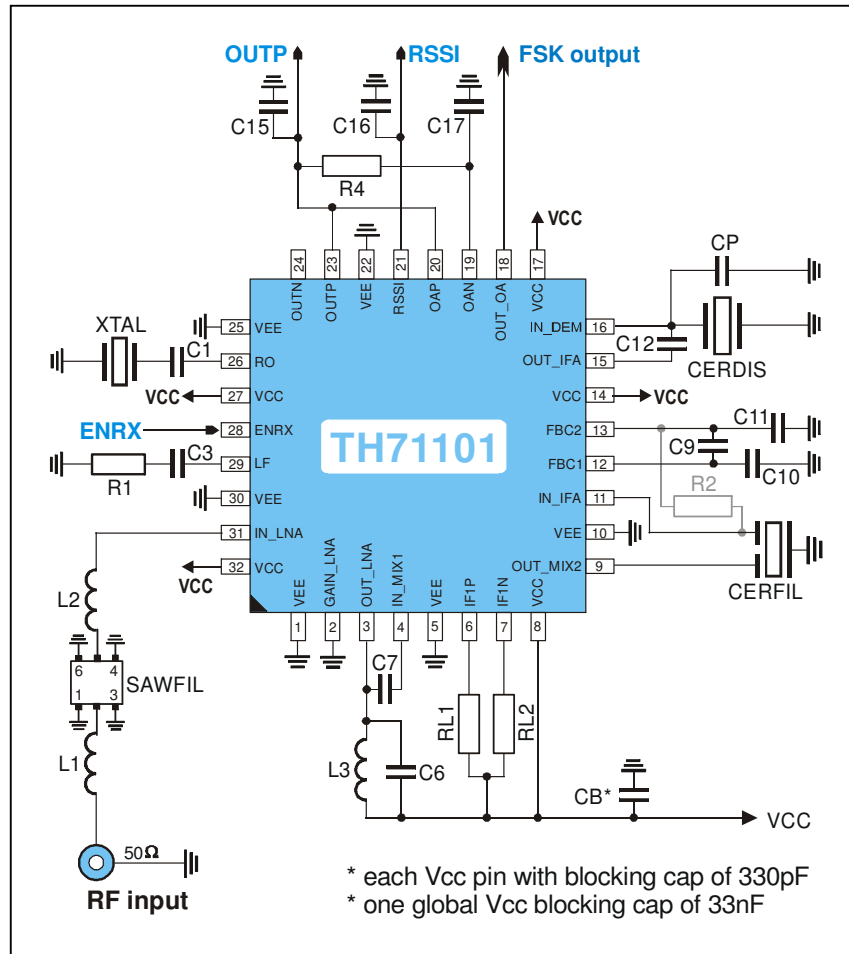


Fig. 3: Test circuit for FSK reception (narrow band)

Circuit Features

- Applicable for narrow band FSK

4.2.2 Narrow Band FSK Component List

Part	Size	Value @ 433.92 MHz	Tolerance	Description
C1	0805	27 pF	±5%	crystal series capacitor
C3	0603	1 nF	±10%	loop filter capacitor
C6	0603	4.7 pF	±5%	LNA output tank capacitor
C7	0603	2.2 pF	±5%	MIX1 input matching capacitor
C9	0603	33 nF	±10%	IFA feedback capacitor
C10	0603	1 nF	±10%	IFA feedback capacitor
C11	0603	1 nF	±10%	IFA feedback capacitor
C12	0805	1.5 pF	±5%	DEMODO phase-shift capacitor
C15	0805	220 pF	±5%	demodulator output low-pass capacitor, this value for data rates < 10 kbps NRZ
C16	0805	1.5 nF	±10%	RSSI output low-pass capacitor
C17	0805	10 nF	±10%	data slicer capacitor, this value for data rates > 0.8 kbps NRZ
CP	0603	6.8 - 8.2 pF	±5%	ceramic resonator loading capacitor
R1	0603	10 kΩ	±5%	loop filter resistor
R2	0603	330 Ω	±5%	optional CERFIL output matching resistor
R4	0805	330 kΩ	±5%	data slicer resistor
RL1	0805	470 Ω	±5%	MIX1 bias resistor
RL2	0805	470 Ω	±5%	MIX1 bias resistor
L1	0603	68 nH	±5%	SAW filter matching inductor from Würth-Elektronik (WE-KI series), or equivalent part
L2	0603	82 nH	±5%	
L3	0603	15 nH	±5%	LNA output tank inductor from Würth-Elektronik (WE-KI series), or equivalent part
XTAL	SMD 6x3.5	26.45125 MHz @ RF = 433.92 MHz	±25ppm cal. ±30ppm temp.	fundamental-mode crystal from Telcona/Horizon or equivalent part
SAWFIL	SMD 3x3	SAFCC433MBL0X00 (f ₀ = 433.92 MHz)	B _{3dB} = 840 kHz	low-loss SAW filter from Murata, or equivalent part
CERFIL	Leaded type	SFKLA10M7NL00	B _{3dB} = 30 kHz	ceramic filter from Murata, or equivalent part
		SFVLA10M7LF00	B _{3dB} = 80 kHz	optional, ceramic filter from Murata, or equivalent part
CERDIS	SMD 4.5x2	CDSCB10M7GA135		ceramic discriminator from Murata, or equivalent part

- For component values for other frequencies, please refer to the EVB descriptions

4.3 ASK Reception

4.3.1 ASK Application Circuit

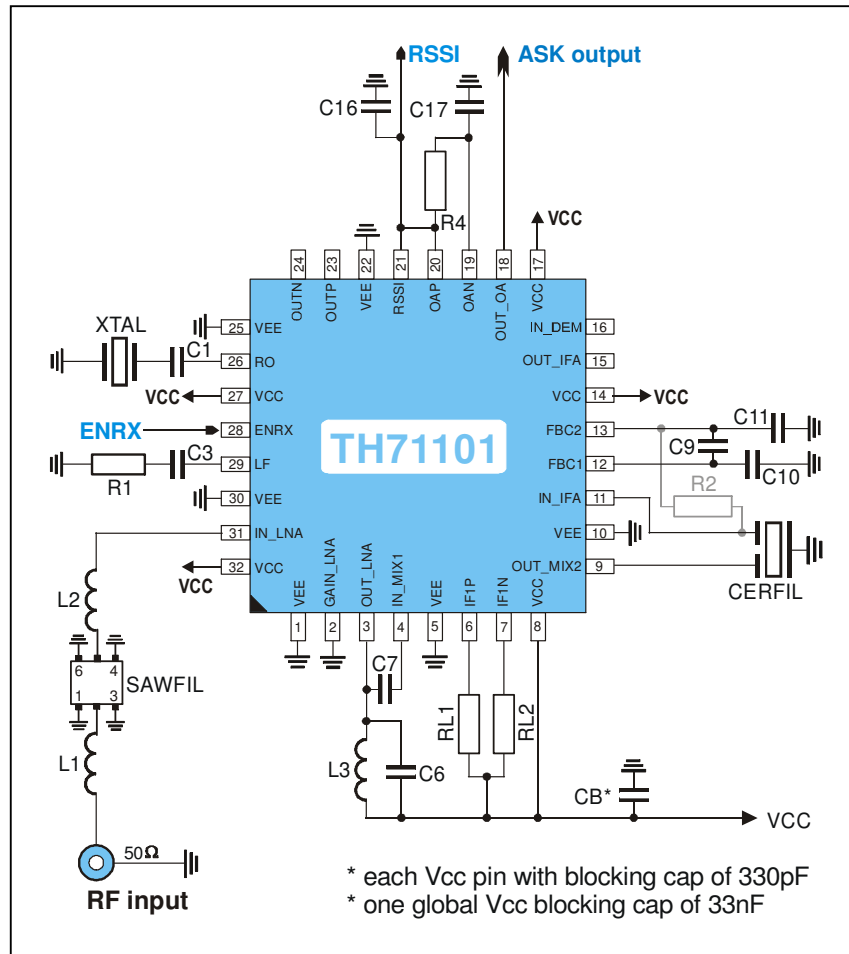



Fig. 5: Test circuit for ASK reception

4.3.2 ASK Component List

Part	Size	Value @ 433.92 MHz	Tolerance	Description
C1	0805	27 pF	±5%	crystal series capacitor
C3	0603	1 nF	±10%	loop filter capacitor
C6	0603	4.7 pF	±5%	LNA output tank capacitor
C7	0603	2.2 pF	±5%	MIX1 input matching capacitor
C9	0603	33 nF	±10%	IFA feedback capacitor
C10	0603	1 nF	±10%	IFA feedback capacitor
C11	0603	1 nF	±10%	IFA feedback capacitor
C16	0805	1.5 nF	±10%	RSSI output low-pass capacitor, this value for data rates < 10 kbps NRZ
C17	0805	10 nF	±10%	data slicer capacitor, this value for data rates > 0.8 kbps NRZ
R1	0603	10 kΩ	±5%	loop filter resistor
R2	0603	330 Ω	±5%	optional CERFIL output matching resistor
R4	0805	330 kΩ	±5%	data slicer resistor
RL1	0805	470 Ω	±5%	MIX1 bias resistor
RL2	0805	470 Ω	±5%	MIX1 bias resistor
L1	0603	68 nH	±5%	SAW filter matching inductor from Würth-Elektronik (WE-KI series), or equivalent part
L2	0603	82 nH	±5%	
L3	0603	15 nH	±5%	LNA output tank inductor from Würth-Elektronik (WE-KI series), or equivalent part
XTAL	SMD 6x3.5	26.45125 MHz @ RF = 433.92 MHz	±25ppm cal. ±30ppm temp.	fundamental-mode crystal from Telcona/Horizon or equivalent part
SAWFIL	SMD 3x3	SAFCC433MBL0X00 (f ₀ = 433.92 MHz)	B _{3dB} = 840 kHz	low-loss SAW filter from Murata, or equivalent part
CERFIL	SMD 3.45x3.1	SFECF10M7HA00	B _{3dB} = 180 kHz	ceramic filter from Murata, or equivalent part
	Leaded type	SFVLA10M7LF00	B _{3dB} = 80 kHz	optional, ceramic filter from Murata, or equivalent part

- For component values for other frequencies, please refer to the EVB descriptions

5 Package Description

 The device TH71101 is RoHS compliant.

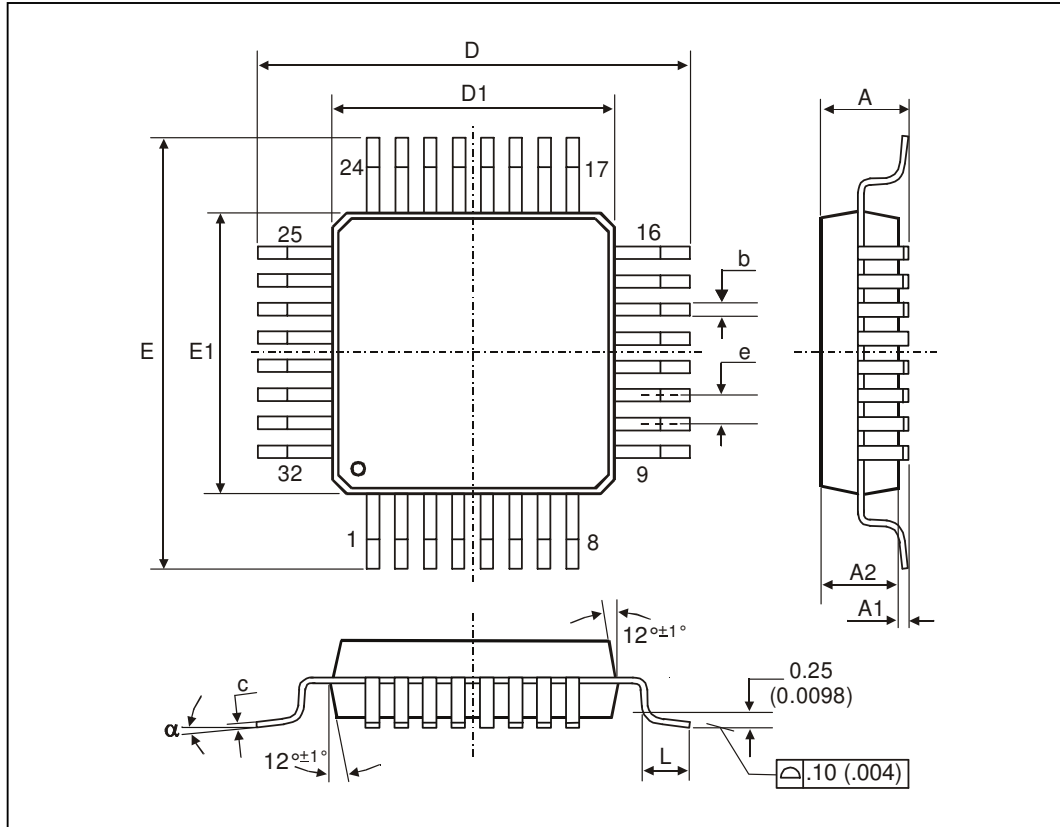


Fig. 6: LQFP32 (Low profile Quad Flat Package)

All Dimension in mm, coplanarity < 0.1mm										
	E1, D1	E, D	A	A1	A2	e	b	c	L	α
min	7.00	9.00	1.40	0.05	1.35	0.8	0.30	0.09	0.45	0°
max			1.60	0.15	1.45		0.45	0.20	0.75	7°
All Dimension in inch, coplanarity < 0.004"										
min	0.276	0.354	0.055	0.002	0.053	0.031	0.012	0.0035	0.018	0°
max			0.063	0.006	0.057		0.018	0.0079	0.030	7°

5.1 Soldering Information

- The device TH71101 is qualified for MSL3 with soldering peak temperature 260 deg C according to JEDEC J-STD-20.

6 Standard information regarding manufacturability of Melexis products with different soldering processes

Our products are classified and qualified regarding soldering technology, solderability and moisture sensitivity level according to following test methods:

Reflow Soldering SMD's (Surface Mount Devices)

- IPC/JEDEC J-STD-020
Moisture/Reflow Sensitivity Classification for Nonhermetic Solid State Surface Mount Devices (classification reflow profiles according to table 5-2)
- EIA/JEDEC JESD22-A113
Preconditioning of Nonhermetic Surface Mount Devices Prior to Reliability Testing (reflow profiles according to table 2)

Wave Soldering SMD's (Surface Mount Devices) and THD's (Through Hole Devices)

- EN60749-20
Resistance of plastic- encapsulated SMD's to combined effect of moisture and soldering heat
- EIA/JEDEC JESD22-B106 and EN60749-15
Resistance to soldering temperature for through-hole mounted devices

Iron Soldering THD's (Through Hole Devices)

- EN60749-15
Resistance to soldering temperature for through-hole mounted devices

Solderability SMD's (Surface Mount Devices) and THD's (Through Hole Devices)

- EIA/JEDEC JESD22-B102 and EN60749-21
Solderability

For all soldering technologies deviating from above mentioned standard conditions (regarding peak temperature, temperature gradient, temperature profile etc) additional classification and qualification tests have to be agreed upon with Melexis.

The application of Wave Soldering for SMD's is allowed only after consulting Melexis regarding assurance of adhesive strength between device and board.

Melexis is contributing to global environmental conservation by promoting **lead free** solutions. For more information on qualifications of **RoHS** compliant products (RoHS = European directive on the Restriction Of the use of certain Hazardous Substances) please visit the quality page on our website:
<http://www.melexis.com/quality.aspx>

7 ESD Precautions

Electronic semiconductor products are sensitive to Electro Static Discharge (ESD). Always observe Electro Static Discharge control procedures whenever handling semiconductor products.

8 Disclaimer

Devices sold by Melexis are covered by the warranty and patent indemnification provisions appearing in its Term of Sale. Melexis makes no warranty, express, statutory, implied, or by description regarding the information set forth herein or regarding the freedom of the described devices from patent infringement. Melexis reserves the right to change specifications and prices at any time and without notice. Therefore, prior to designing this product into a system, it is necessary to check with Melexis for current information. This product is intended for use in normal commercial applications. Applications requiring extended temperature range, unusual environmental requirements, or high reliability applications, such as military, medical life-support or life-sustaining equipment are specifically not recommended without additional processing by Melexis for each application.

The information furnished by Melexis is believed to be correct and accurate. However, Melexis shall not be liable to recipient or any third party for any damages, including but not limited to personal injury, property damage, loss of profits, loss of use, interrupt of business or indirect, special incidental or consequential damages, of any kind, in connection with or arising out of the furnishing, performance or use of the technical data herein. No obligation or liability to recipient or any third party shall arise or flow out of Melexis' rendering of technical or other services.

© 2013 Melexis NV. All rights reserved.

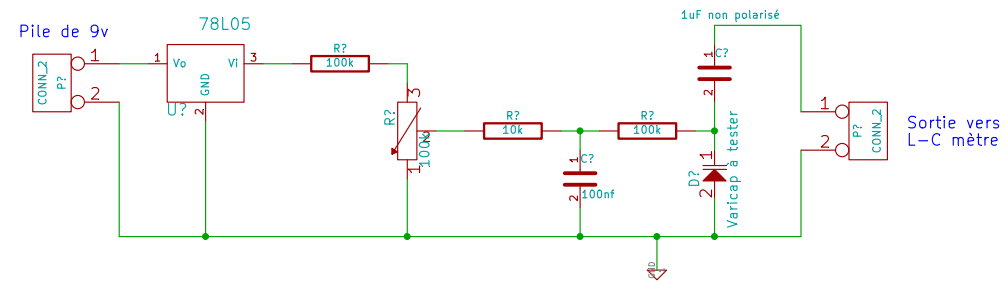
For the latest version of this document, go to our website at
www.melexis.com

Or for additional information contact Melexis Direct:

Europe, Africa, Asia:	America:
Phone: +32 1367 0495	Phone: +1 248 306 5400
E-mail: sales_europe@melexis.com	E-mail: sales_usa@melexis.com

ISO/TS 16949 and ISO14001 Certified

Testeur de diode Varicap



On trouve sur internet des L-C mètre pas cher et qui fonctionne bien

ADRASEC 29		
File: Test Varicap.sch		
Sheet: /		
Title: Testeur Diode Varicap		
Size: A4	Date: 6 jul 2018	Rev: F5LEB
KiCad E.D.A. eeschema (2013-07-07 BZR 4022)-stable		Id: 1/1